
Capabilities and incapacibilities of unity power factor control algorithm for DSTATCOM under abnormal grid and load conditions

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Abstract: Distribution static compensator (DSTATCOM) is an efficient custom power device to enhance the power quality of distribution network. To operate the DSTATCOM, a control strategy is a must and the unity power factor (UPF) is one such control strategy, which is used in DSTATCOM applications. The UPF algorithm has been developed in such a way that it operates as the reference supply current generator for DSTATCOM. To ensure the effective operation of DSTATCOM, it is important to analyse the performance of selected control algorithm at different disturbances either coming from source side or load side. A three-phase four-wire distribution system is considered for this study and the performance of UPF-based DSTATCOM system has been analysed for different disturbances such as unbalanced and harmonic distortion coming from source and load side. Exhaustive analysis is provided using the simulation tools (MATLAB/Simulink) to show the capabilities and incapacibilities of UPF control algorithm under such disturbances. Each case is analysed thoroughly keeping the view of reactive power requirement, load balancing, total harmonic distortion levels as per IEEE-519 standard and power factor improvement.

Keywords: current distortion; distribution static compensator; DSTATCOM; unbalancing; unity power factor; UPF; control algorithm; voltage distortion.

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1 Introduction

Use of power electronic converters (PECs) is continuously growing in domestic, commercial and industrial applications to provide operating flexibility and energy efficiency (Bose, 2009). Therefore, these are very common and present in most of the applications. These power electronic-based loads are nonlinear in nature and therefore possess some disadvantages such as reactive power requirement and generation of harmonics which lead to poor power factor (PF). Moreover, these current harmonics together with grid impedance results voltage harmonic distortion at the point of common coupling (PCC) (Kumar and Zare, 2016; Zare and Kumar, 2016; Zare et al., 2017). Modern distribution network faces a variety of power quality issues such as voltage unbalance, background harmonic distortion, voltage sag, swell and so on. Recent power quality surveys show that voltage unbalance is one of major problem in modern distribution network (CEER, 2011; Brito et al., 2018; Elphick et al., 2017; Kumar et al., 2017). Root causes of voltage unbalance in a distribution network are (Von Jouanne and Banerjee, 2001):

- 1 unequal distribution of single-phase load on three-phase power system
- 2 asymmetrical feeders and transformer winding
- 3 asymmetrical power generation of grid connected single phase distributed generations such as roof-top solar inverters.

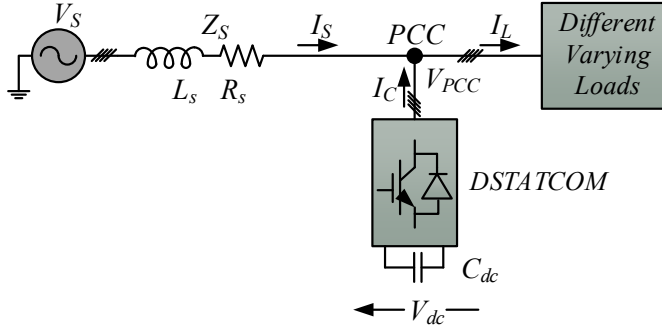
The use of three-phase four-wire (3P4W) configuration is common in different part of world to supply power to residential and commercial loads. The unbalance in such distribution network can overload neutral wire due to high current flow (Alam et al., 2015).

In order to improve the power quality at PCC, a number of solutions have been proposed such as passive filters, active filters, hybrid filter, etc. Passive filters are normally tuned at specific frequency. Therefore, many tuned passive filters are required in a complex system (Albatran et al., 2018; Aleem and Zobaa, 2017; Kumar and Zare, 2014; Leite et al., 2017). Moreover, passive filters have some severe drawbacks such as aging, large parameter tolerance and resonance issue with grid impedance (Das, 2003; Menti et al., 2009). Also, these passive filters will not help to improve voltage unbalancing issue. Active filters are best solution for handling many problems at a time (Akagi, 2005; Litrán et al., 2005; Rohilla and Pal, 2013; Singh et al., 1999) but their cost is relatively higher than passive filters, which eliminates them to use in many cost-sensitive applications. To minimise this drawback, various hybrid filters topologies have been reported in recent years (Asiminoaei et al., 2006; Chen and von Jouanne, 2001; George and Mishra, 2010; Gupta et al., 2011; Rahmani et al., 2009). Hybrid filters are combination of active and passive filters and could provide a cost-effective solution.

The distribution static compensator (DSTATCOM) is an example of shunt connected active filter and has the functions of reactive power compensation and harmonic mitigation. Single line diagram of DSTATCOM is shown in Figure 1. The DSTATCOM for 3P4W distribution system can be formed using different configurations such as three-leg voltage source converter (VSC) with single DC capacitor, three-leg VSC with neutral clamped DC capacitors, three-leg VSC with transformer, four-leg VSC, etc. (Iyer et al., 2005; Kumar et al., 2008; Mahela and Shaik, 2015; Srikanthan and Mishra, 2009).

The main purpose of DSTATCOM is to supply reactive power demanded by load, so that source only supply active power and results in unity power factor (UPF). If load is nonlinear in nature, then the load current (I_L) has both active and reactive power components. Active power requirement of load current is fulfilled by I_S (source current) and reactive power requirement is fulfilled by I_C (compensator current) as shown in Figure 1.

Figure 1 Single line diagram of DSTATCOM (see online version for colours)



Many control algorithms are available to enable DSTATCOM to serve as a reactive power compensator at PCC (Bangarraju et al., 2014; Benysek and Pasko, 2012; Elhaj et al., 2018; Han et al., 2010; Krismanto et al., 2012; Meral and Çelik, 2018; Saralaya and Sharma, 2018). The UPF is one such algorithm (Abellan et al., 2002; Garg et al., 2014; Orts-Grau et al., 2008). Orts-Grau et al. (2008) analysed UPF control algorithm for 3P4W system. Garg et al. (2014) shows the UPF control theory performance for the induction motor fed from single-phase supply and Abellan et al. (2002) used this control algorithm for three-phase supply system. These papers analysed the UPF performance in the presence of load side disturbances. Impact of source/grid side disturbances is still unknown for this control algorithm which needs to be explored to have complete performance study. In order to fill this gap, this paper analysed the performance of UPF-based DSTATCOM control algorithm in detail and observed its capabilities and incapacities at different grid and load profiles.

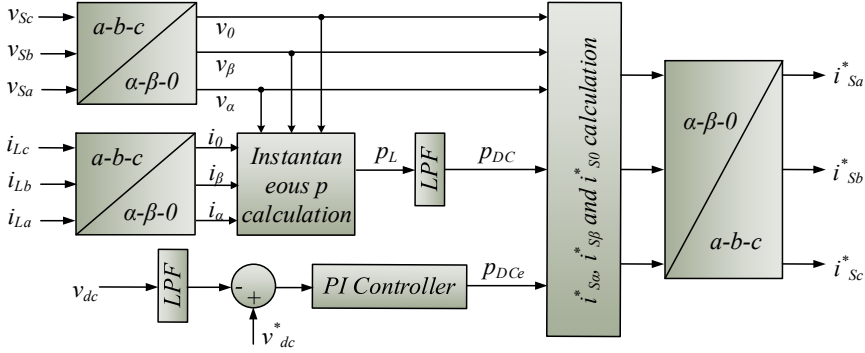
The rest of this paper is organised as follows. In Section 2, the UPF control algorithm is described analytically. In Section 3, performance of UPF is analysed under different load and grid conditions and obtained simulation results are discussed in detail. Section 4 summarises the performance of UPF control algorithm. Finally, conclusion is drawn in Section 5.

2 UPF control algorithm

As the name suggests the UPF control algorithm is developed to provide UPF in normal or abnormal load conditions. Any system will supply UPF if it has a resistive load. In other words, system will provide UPF if its voltage space vector is in phase with current space vector. The operating principle of UPF is to see load as a resistive component at PCC. Purpose of this control algorithms is to generate reference source currents, so that source should supply only active power and rest power requirement of load is supplied by

compensator, i.e., DSTATCOM in this study. So, reference source currents are corresponding to load but in phase with source voltage. The detail block diagram for extracting reference source currents using UPF control algorithm is shown in Figure 2.

Figure 2 Block diagram for extracting reference source currents using UPF control algorithm (see online version for colours)



The UPF method is also known as ‘voltage synchronisation method’ where source current space vector is desired to be in phase with the PCC voltage space vector as shown in equation (1):

$$i_s^* = K \cdot v \quad (1)$$

where K is a constant. Value of K depends on the PCC voltages (v) and the load. The i_s^* is the required reference source currents. The power delivered by the source will be

$$p_s = v \cdot i_s^* = K \cdot v^2 \quad (2)$$

v can be given in α - β -0 plane as

$$v = \sqrt{v_0^2 + v_\alpha^2 + v_\beta^2}$$

To calculate p_s , load currents (i.e., i_{La} , i_{Lb} and i_{Lc}) and PCC voltages (i.e., v_{Sa} , v_{Sb} and v_{Sc}) are sensed and converted into α - β -0 coordinates as shown in equations (3) and (4).

$$\begin{bmatrix} v_\alpha \\ v_\beta \\ v_0 \end{bmatrix} = \sqrt{\frac{2}{3}} \begin{bmatrix} 1 & -1/2 & -1/2 \\ 0 & \sqrt{3}/2 & -\sqrt{3}/2 \\ 1/\sqrt{2} & 1/\sqrt{2} & 1/\sqrt{2} \end{bmatrix} \cdot \begin{bmatrix} v_{Sa} \\ v_{Sb} \\ v_{Sc} \end{bmatrix} \quad (3)$$

$$\begin{bmatrix} i_\alpha \\ i_\beta \\ i_0 \end{bmatrix} = \sqrt{\frac{2}{3}} \begin{bmatrix} 1 & -1/2 & -1/2 \\ 0 & \sqrt{3}/2 & -\sqrt{3}/2 \\ 1/\sqrt{2} & 1/\sqrt{2} & 1/\sqrt{2} \end{bmatrix} \cdot \begin{bmatrix} i_{La} \\ i_{Lb} \\ i_{Lc} \end{bmatrix} \quad (4)$$

The constant K can be determined with the criterion that the power delivered by the source equal to the DC component of the instantaneous active power of the load, so that

$$K = \frac{p_{DC} + p_{DCe}}{(v_0^2 + v_\alpha^2 + v_\beta^2)_{DC}} \quad (5)$$

p_{DC} and p_{DCe} are explained in equations (7) and (8), respectively. The instantaneous active power in α - β -0 coordinates are calculated as:

$$p_L = v_\alpha \cdot i_\alpha + v_\beta \cdot i_\beta + v_0 \cdot i_0 \quad (6)$$

Total instantaneous active power (p_L) can be decomposed into fundamental and harmonic powers as:

$$p_L = \bar{p}_L + \tilde{p}_L = p_{DC} + p_{AC} \quad (7)$$

where p_{DC} represents the DC component which is responsible for fundamental active power and p_{AC} represents harmonic power component. As source is required to supply fundamental power (p_{DC}) only and the other power requirement, i.e., reactive power (p_{AC}) must be supplied by DSTATCOM. The fundamental instantaneous active power component can be extracted from the p_L by using a low pass filter (Butterworth type).

The reference source currents in α - β -0 reference frame can be calculated as:

$$\begin{bmatrix} i_{S\alpha}^* \\ i_{S\beta}^* \\ i_{S0}^* \end{bmatrix} = \frac{1}{Ax_{(3\phi)}} \cdot \begin{bmatrix} v_\alpha \\ v_\beta \\ v_0 \end{bmatrix} \cdot [p_{DC} + p_{DCe}] \quad (8)$$

where $i_{S\alpha}^*$, $i_{S\beta}^*$ and i_{S0}^* represent the reference source current signals in α - β -0 coordinates respectively. $Ax_{(3\phi)} = (v_0^2 + v_\alpha^2 + v_\beta^2)_{DC}$ and p_{DCe} is active power required to maintain the DC bus voltage at a constant reference value (V_{DC}^*). The p_{DCe} is the losses of DSTATCOM. Aim of this paper is to analyse the capabilities and incapacibilities of UPF control algorithm in extracting reference source currents. For doing so, there is no need to calculate p_{DCe} . Good PI tuned DSTATCOM follows nearly same output characteristics as generated by control algorithms (Bangarraju et al., 2014; Orts-Grau et al., 2008; Rohilla and Pal, 2013).

Reference source currents in a - b - c reference frame can be calculated by inverse transformation of equation (8) as:

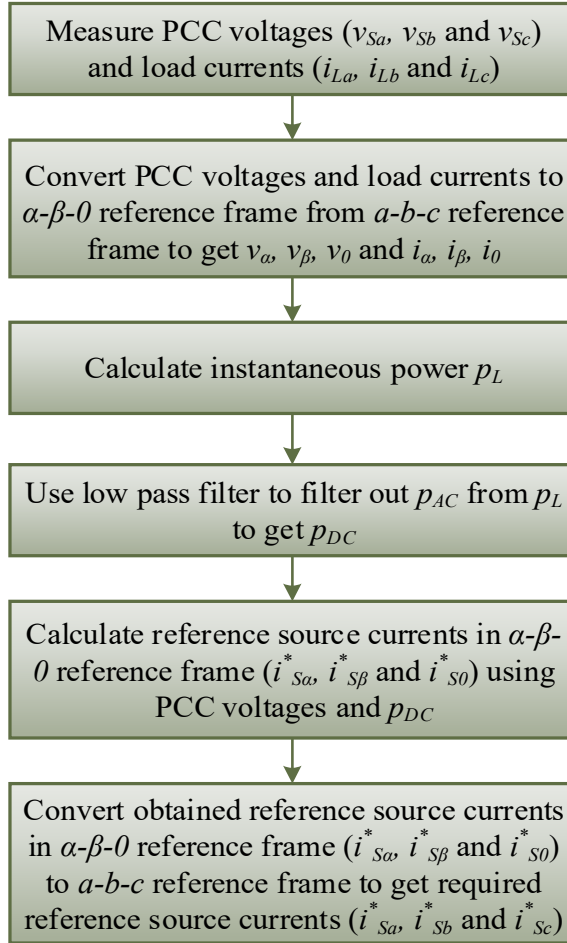
$$\begin{bmatrix} i_{Sa}^* \\ i_{Sb}^* \\ i_{Sc}^* \end{bmatrix} = \sqrt{\frac{2}{3}} \cdot \begin{bmatrix} 1 & 0 & 1/\sqrt{2} \\ -1/2 & \sqrt{3}/2 & 1/\sqrt{2} \\ -1/2 & -\sqrt{3}/2 & 1/\sqrt{2} \end{bmatrix} \cdot \begin{bmatrix} i_{S\alpha}^* \\ i_{S\beta}^* \\ i_{S0}^* \end{bmatrix} \quad (9)$$

where i_{Sa}^* , i_{Sb}^* and i_{Sc}^* are the required reference source currents for phase a , phase b and phase c , respectively. These extracted reference source currents (i_{Sa}^* , i_{Sb}^* and i_{Sc}^*) used to generate switching signals of the DSTATCOM.

Step by step flow diagram to extract reference source currents using UPF control algorithm is shown in Figure 3. First step is to measure PCC voltages and load currents.

In second step, these measured quantities are converted into α - β -0 reference frame from a - b - c reference frame. In third step, instantaneous power is calculated. In next step, this calculated power is filtered to find out DC component of it. DC component of power and α - β -0 reference voltages are used to calculate α - β -0 values of reference source currents in next step. In last step, α - β -0 values are converted to a - b - c values to obtain required reference source currents.

Figure 3 Flow diagram of UPF control algorithm (see online version for colours)



3 Performance evaluation of UPF control theory

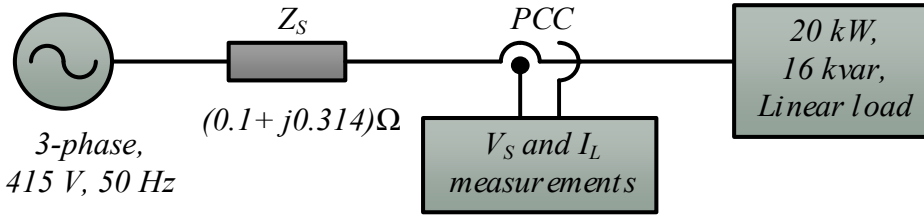
In this section, performance of UPF control algorithm is analysed at different load and supply conditions as illustrated in Table 1. Simulation tool of matrix laboratory has been used for this purpose. In all the cases, a 3P4W of 415 V line-line, 50 Hz and a linear load of 20 kW with 0.8 PF has been considered in general, as shown in Figure 4. Line impedance (Z_S) of $(0.14 + j0.314) \Omega$ has been considered for this study.

Table 1 Different cases considered for UPF evaluation

Cases	Source		Load	
	Specifications	Distortion	Specifications	Distortion
Case I	Ideal supply voltage – balanced and undistorted load current	415 V L-L, 50 Hz, $Z_S = (0.14 + j0.314) \Omega$	20 kW, 0.8 PF	No distortion
Case II	Ideal supply voltage – balanced and distorted load current (5th, 7th, 11th harmonics)	415 V L-L, 50 Hz, $Z_S = (0.14 + j0.314) \Omega$	20 kW, 0.8 PF	$THD_i = 16.08\%$
Case III	Distorted supply voltage (5th harmonic) – balanced and undistorted load current	415 V L-L, 50 Hz, $Z_S = (0.14 + j0.314) \Omega$, 5th harmonic = $10\% V_1$	20 kW, 0.8 PF	No distortion
Case IV	Distorted supply voltage (5th harmonic) – balanced and distorted load current (5th, 7th, 11th harmonics)	415 V L-L, 50 Hz, $Z_S = (0.14 + j0.314) \Omega$ 5th harmonic = $10\% V_1$	20 kW, 0.8 PF	$THD_i = 16.08\%$
Case V	Ideal supply voltage – unbalanced and undistorted load current (1-phase out)	415 V L-L, 50 Hz, $Z_S = (0.14 + j0.314) \Omega$ 5th harmonic = $10\% V_1$	20 kW, 0.8 PF, phase 'b' out	Load unbalancing
Case VI	Distorted supply voltage (5th harmonic) – unbalanced and undistorted load current	415 V L-L, 50 Hz, $Z_S = (0.14 + j0.314) \Omega$, 5th harmonic = $10\% V_1$	20 kW, 0.8 PF, additional 5 Ω in phase a	Load unbalancing
Case VII	Distorted supply voltage (5th harmonic) – unbalanced and distorted load current (5th, 7th, 11th harmonics)	415 V L-L, 50 Hz, $Z_S = (0.14 + j0.314) \Omega$ 5th harmonic = $10\% V_1$	20 kW, 0.8 PF, additional 5 Ω in phase a	$THD_i = 16.08\%$
Case VIII	Unbalanced supply voltage – balanced and undistorted load current	415 V L-L, 50 Hz, $Z_S = (0.14 + j0.314) \Omega$, additional 5 Ω in phase a	20 kW, 0.8 PF	No distortion

Note: V_1 = Magnitude of fundamental component of source voltage.

Figure 4 Single line diagram of system under consideration (see online version for colours)

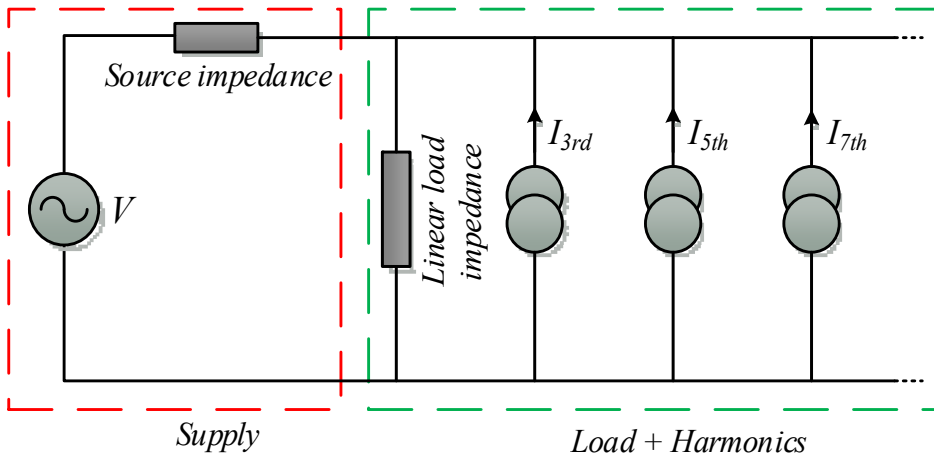


To simplify the simulation model and reduce the execution time, only linear load has been considered. While still retaining the nonlinearity in the system, harmonics are inserted in the linear loads as shown in Figure 5 (Chapman, 2001). Distortion in source, load and in reference source currents are measured by calculating total harmonic distortion (*THD*) as given in equation (10). Harmonics have been measured up to 20th order.

$$THD_x = \frac{\sqrt{x_2^2 + x_3^2 + \dots + x_{20}^2}}{x_1} \tag{10}$$

where *x* represents current or voltage for which distortion need to be measured. The x_1, x_2, \dots, x_{20} are the RMS values of *x* for 1st, 2nd, ..., 20th order harmonics, respectively.

Figure 5 Equivalent circuit of a nonlinear load (see online version for colours)



3.1 Case I: ideal supply voltage – balanced and undistorted load current

For benchmarking study, balanced and undistorted grid and load conditions have been considered in this case. A 20 kW linear load is connected and PF has been kept at 0.8. Figure 6(a) shows the supply voltages (v_s), load currents (i_L) and reference supply currents (i_s^*). As no distortion and no unbalancing are considered, so all the three waveforms observed as sinusoidal in nature.

Figure 6 Simulation results for Case I, (a) supply voltages, load currents, and reference source currents (b) $0-\alpha-\beta$ coordinates of supply voltages (c) harmonic analysis of phase *a* of load current and reference source current (d) Source voltage and current, source voltage and reference current of phase *a* for PF analysis (see online version for colours)

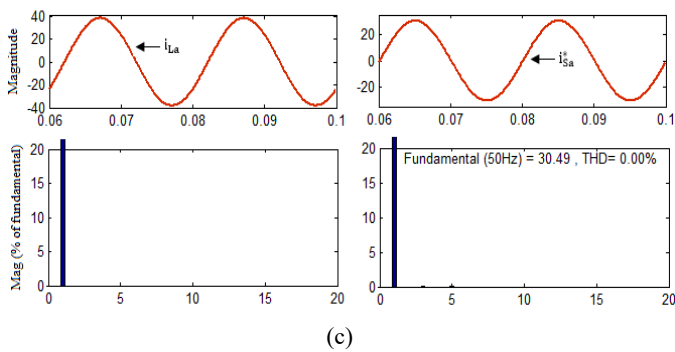
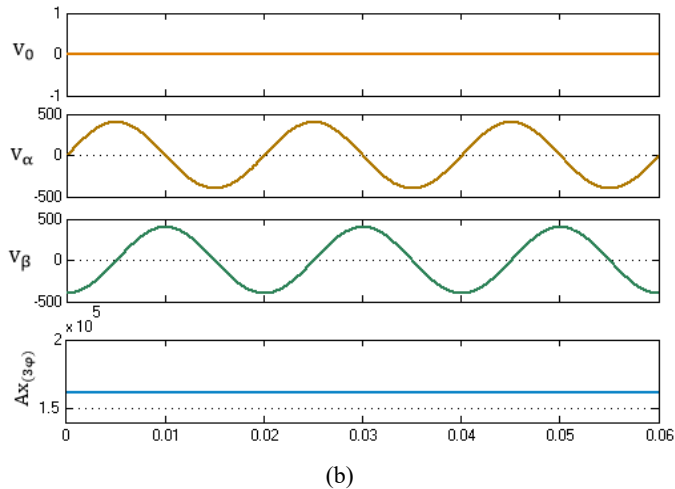
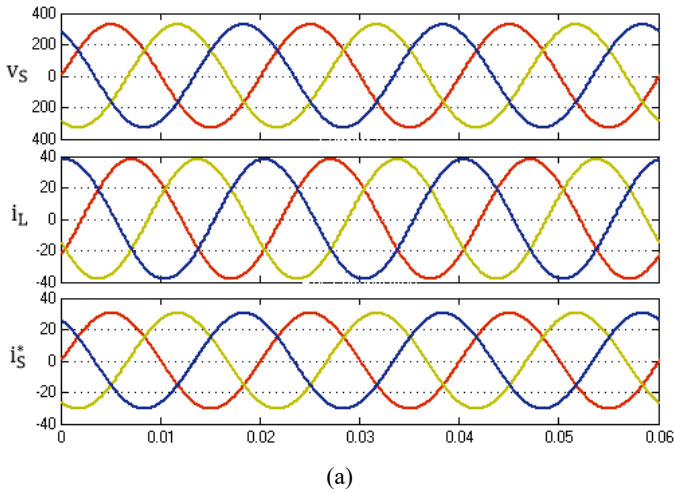


Figure 6 Simulation results for Case I, (a) supply voltages, load currents, and reference source currents (b) 0- α - β coordinates of supply voltages (c) harmonic analysis of phase *a* of load current and reference source current (d) Source voltage and current, source voltage and reference current of phase *a* for PF analysis (continued) (see online version for colours)

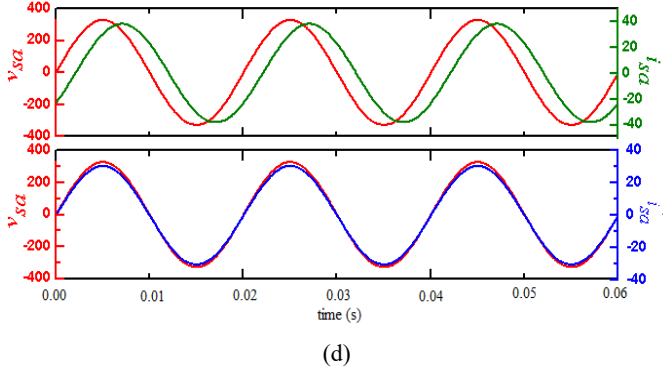


Figure 6(c) represents the harmonic analysis of load current and reference source current at PCC. Only phase *a* harmonic analysis is shown here, as harmonics are same in other two phases *b* and *c* unless system is unbalanced.

Figure 6(b) shows the four waveforms in which first three are v_0 , v_α and v_β waveforms which are 0- α - β reference frame converted from *a-b-c* reference frame and fourth waveform shows $Ax_{(3\phi)}$. As observed from Figure 6(b), v_0 is zero and v_α and v_β are 90° phase shifted.

Figure 6(d) depicts the capability of PF correction of the UPF control algorithm. In first waveform of Figure 6(d), the load has 0.8 PF lagging which can be seen from the phase difference between the waveforms of $v_{S\alpha}$ and $i_{S\alpha}$. This phase difference came down to zero as shown by $v_{S\alpha}$ and $i_{S\alpha}$ as a result, the PF has corrected to unity as shown in second waveform. Simulation results are summarised in Table 2.

Table 2 Simulation results summary of Case I

	i_{La}	i_{Sa}^*
THD_i	0.00	0.00
I (A)	26.95	21.56
P (W)	5,003	5,003
Q (var)	3,752	0.038
PF	0.8	1.0
$THD_{va} = 0.00$		

In Table 2, column i_{La} and column i_{Sa}^* show the simulation results obtained for load current for phase *a* and reference source (grid) current for phase *a* respectively, generated by the UPF control algorithm. The I , P and Q are current, active power and reactive power respectively. In this case, the THD is zero, because balanced and undistorted supply and load conditions have been considered in this case. Here, it is important to note that the PF has been improved from 0.8 to 1.0, which shows the effectiveness of UPF

control algorithm. Active power (P) supplied by source is constant at 5,003 watts. Reactive power (Q) requirement of load is 3,752 vars. This reactive power is expected to supply by compensator (i.e., DSTATCOM), which is confirmed by results presented in Table 2 where UPF has successfully compensated this reactive power requirement and reduced it to 0.038 vars. Now, as the reactive power requirement has been compensated, so reference source current is reduced to 21.56 A in comparison to load requirement of 26.95 A.

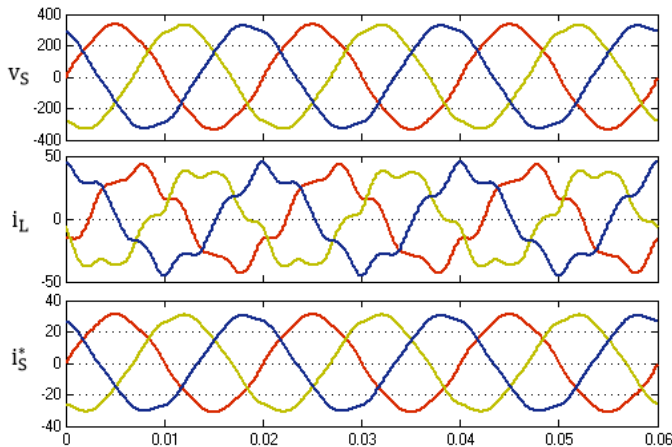
‘From Figure 6 and Table 2, one can conclude that UPF control theory is capable to improve PF by compensating reactive power required by load in case of ideal supply voltage, balanced and undistorted load condition.’

3.2 Case II: ideal supply voltage – balanced and distorted load current (5th, 7th, 11th harmonics)

Supply voltage conditions are same as in Case I. In the load 5th, 7th and 11th harmonics are inserted (as per Figure 5) to analyse the performance of UPF control theory. Concerned waveforms are shown in Figure 7.

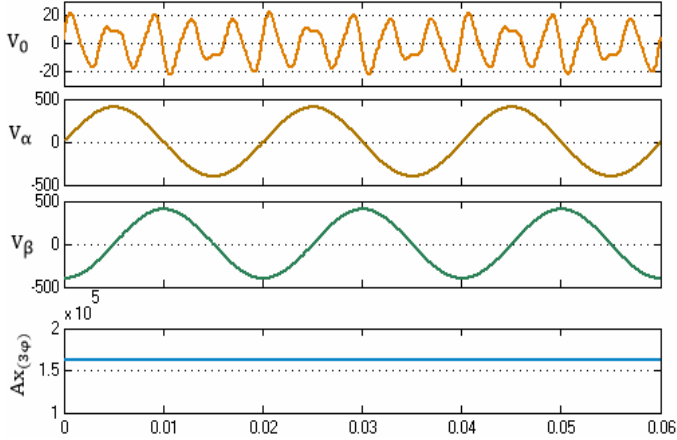
Figure 7(a) shows distortion in load current i_L due to presence of 5th, 7th and 11th harmonics. This distortion gives 16.08% THD_i in load current as shown in Figure 7(c). This distorted current passes through the source impedance, as a result source voltage gets distorted [as in Figure (a)] and around 3.04% of THD_v has been recorded at PCC. Distortion in load currents has compensated by UPF at PCC and results only 3.03% THD_i in source current (i_s^*) at PCC as shown in Figure 7(c). In Figure 7(b), v_0 shows the effect of distortion on 0- α - β reference frame whereas there is no effect on v_α and v_β , and both has same nature as in Case I. In Figure 7(d), effectiveness of UPF algorithm can be observed where PF is improved from 0.8 to unity.

Figure 7 Simulation results for Case II, (a) supply voltages, load currents, and reference source currents (b) 0- α - β coordinates of supply voltages (c) harmonic analysis of phase a of load current and reference source current (d) source voltage and current, source voltage and reference current of phase a for PF analysis (see online version for colours)

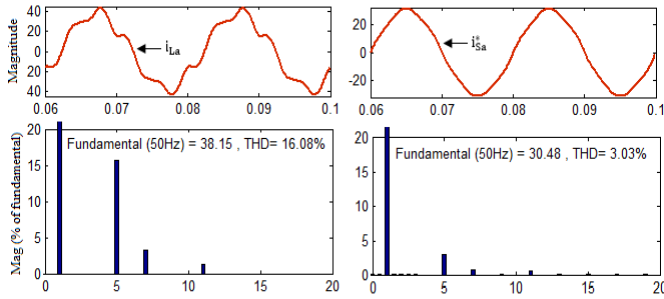


(a)

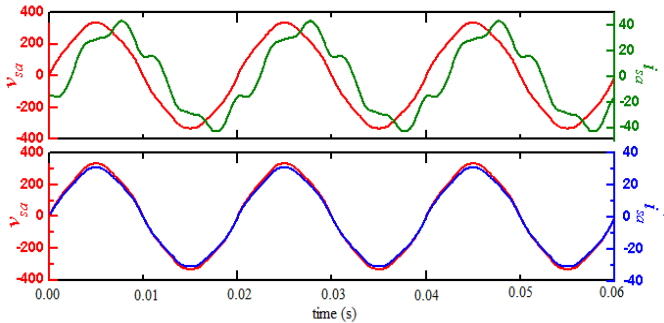
Figure 7 Simulation results for Case II, (a) supply voltages, load currents, and reference source currents (b) $0-\alpha-\beta$ coordinates of supply voltages (c) harmonic analysis of phase a of load current and reference source current (d) source voltage and current, source voltage and reference current of phase a for PF analysis (continued) (see online version for colours)



(b)



(c)



(d)

Various parameters of Case II are shown in Table 3. The UPF has maintained the UPF in this case also by compensating reactive power requirement of 3,760 vars to 0.432 vars which is nearly zero. As a result, reference source current (i_{Sa}^*) is decreased to

21.55 A in comparison to load current requirement of 26.98 A and remaining 5.43 A reactive current component supplied by DSTATCOM.

Table 3 Simulation results summary of Case II

	i_{La}	i_{Sa}^*
THD_i	16.08	3.03
I (A)	26.98	21.55
P (W)	5,013	5,007
Q (var)	3,760	0.432
PF	0.8	1.0
$THD_{va} = 3.04$		

Here, it is interesting to note that compensated current and v_{Sa} , both are showing nearly same distortion (i.e., 3%). It seems that UPF has replicated the THD of source voltage in the THD of reference source current. This is due to fact that UPF has the property to see everything connected after PCC as resistive components but not before PCC.

‘It can be concluded that UPF has compensated harmonics in load current and provided UPF at PCC.’

3.3 Case III: distorted supply voltage (5th harmonic) – balanced and undistorted load current

The 5th harmonic distortion is introduced in supply voltage which results a THD_v of 9.75%. Load is considered same as in Case I, i.e., balanced and undistorted. Its effect on all the other parameters is shown in Figure 8.

Figure 8 Simulation results for Case III, (a) supply voltages, load currents, and reference source currents (b) $0-\alpha-\beta$ coordinates of supply voltages (c) harmonic analysis of phase a of load current and reference source current (d) source voltage and current, source voltage and reference current of phase a for PF analysis (see online version for colours)

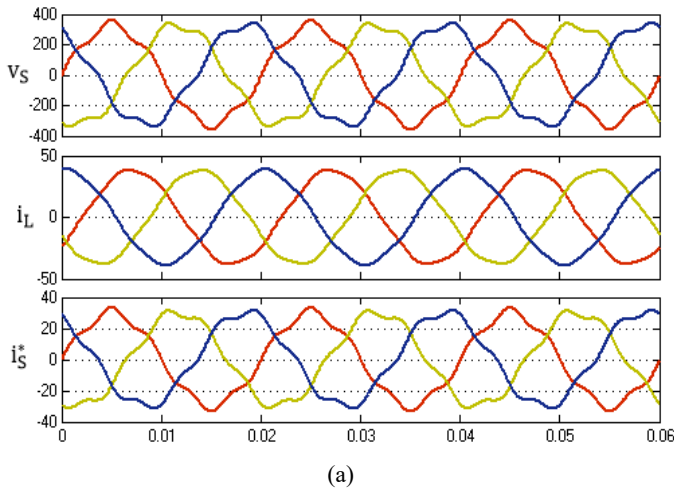
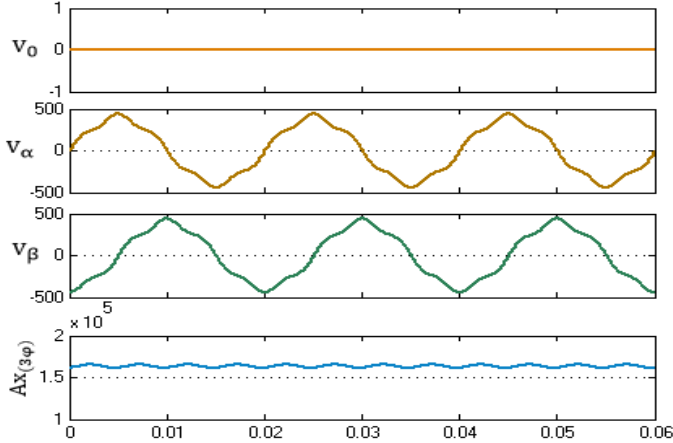
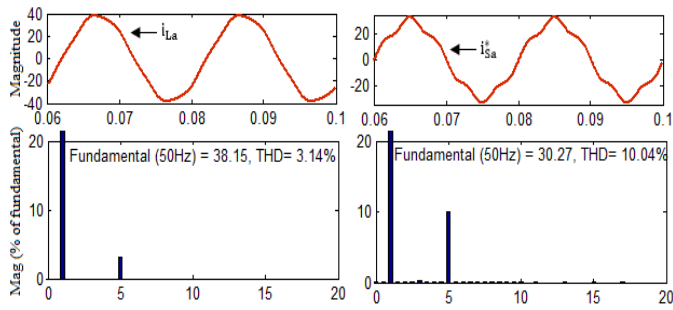


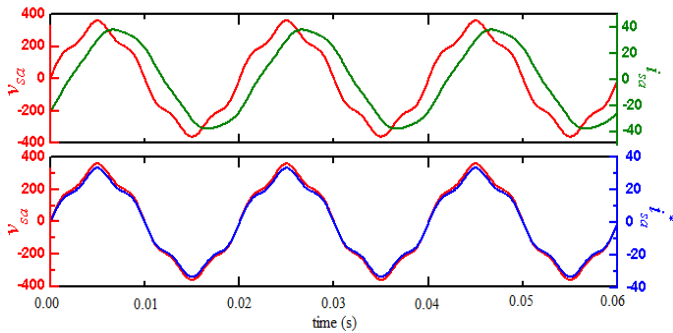
Figure 8 Simulation results for Case III, (a) supply voltages, load currents, and reference source currents (b) $0-\alpha-\beta$ coordinates of supply voltages (c) harmonic analysis of phase a of load current and reference source current (d) source voltage and current, source voltage and reference current of phase a for PF analysis (continued) (see online version for colours)



(b)



(c)



(d)

It can be seen in Figure 8(a), that v_s and i_s^* have same waveform pattern whereas i_L has different pattern. ‘It leads to the conclusion that harmonics of source is also present in reference source current.’ This can also be confirmed by the data shown in Table 4.

Table 4 Simulation results summary of Case III

	i_{La}	i_{Sa}^*
THD_i	3.14	10.04
I (A)	26.98	21.4
P (W)	5,013	4,972
Q (var)	3,760	0.078
PF	0.8	1.0
$THD_{va} = 9.75$		

Voltage distortion does not affect v_0 but effect v_α and v_β , as in Figure 8(b). The i_s^* has a THD_i of 10.04% which is resultant of source voltage THD_v , plus other compensation losses. Due to distorted supply voltage, load current also shows some distortion, 3.14% as THD_i . ‘The UPF has perfectly compensated the reactive power requirement of load and has given a UPF but has no improvement on background distortion in supply side.’

Also, from Figure 7(b) and Figure 8(b), it is observed that distortion in load shows its effect on v_0 but not on v_α and v_β and distortion in source disturb the v_α and v_β but no effect on v_0 . Therefore, by observing the v_0 , v_α and v_β disturbances, it is possible to identify the origin of disturbance from source side or load side.

3.4 Case IV: distorted supply voltage (5th harmonic) – balanced and distorted load current (5th, 7th, 11th harmonics)

In this case, both supply voltage and load current have distortion. Level of distortion is same as considered in Case II (load current THD_i is 16.08%) and Case III (background supply voltage distortion THD_v is 9.75%). Concerned simulation results are shown in Figure 9.

Source voltage and reference source current [Figure 9(d)] are in same phase which results UPF although both waveforms are still distorted. v_0 shows the same nature as was in Case II for distorted load conditions and v_α and v_β show the same nature as was in Case III for distorted load conditions. Load has a THD_i of 13.04% and reference source current has 12.9% as THD_i , replicating the voltage THD_v coming from grid side. Reactive current compensation is up to the mark.

Here, one can observe that applied load current THD_i was 16.08% whereas Table 5 shows 13.04% only. Similarly applied background supply voltage has THD_v 9.75% but Table 5 is showing 12.62%. Reason of this change can be understood by revisiting Case II and Case III, as this case is a summation of both the above cases’ distortions. Case II has a voltage THD_v of 3.04% despite having no distortion in source; it is due to load harmonics interaction with source impedance as stated earlier. Similar behaviour noticed here, when applied distortion of 9.75% in voltages has increased to 12.62%. Now in Case III, load has no distortion but it showed a THD_i of 3.14% due to voltage distortion. Here, it seems that this THD_i has been subtracted from the load and load current THD_i reduced to 13.04% from 16.08% and this is mainly due to background distortion.

Figure 9 Simulation results for Case IV, (a) supply voltages, load currents, and reference source currents (b) 0- α - β coordinates of supply voltages (c) harmonic analysis of phase *a* of load current and reference source current (d) source voltage and current, source voltage and reference current of phase *a* for PF analysis (see online version for colours)

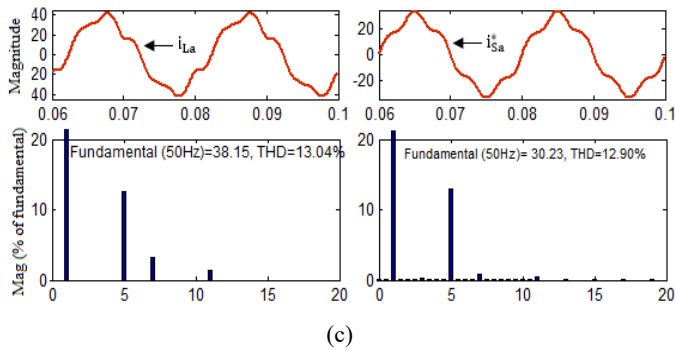
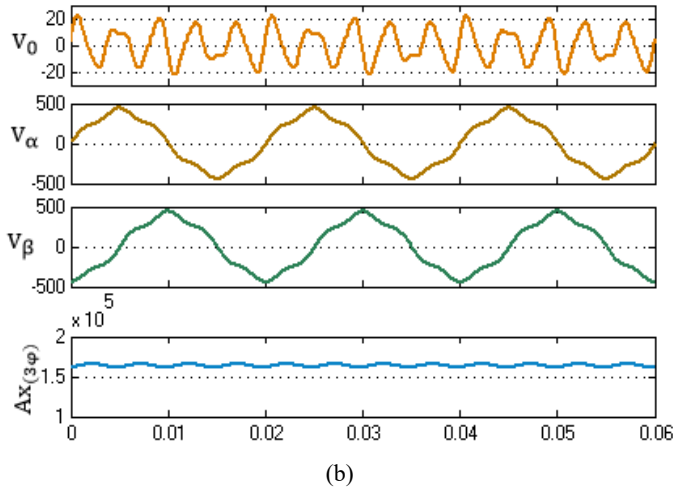
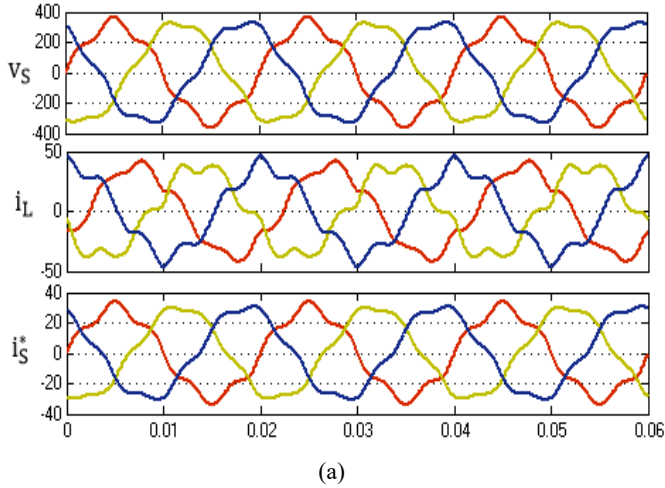
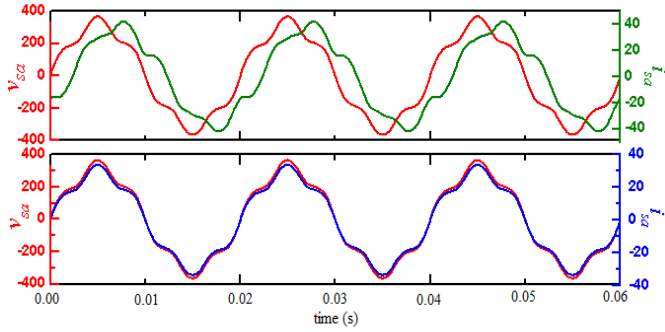


Figure 9 Simulation results for Case IV, (a) supply voltages, load currents, and reference source currents (b) $0-\alpha-\beta$ coordinates of supply voltages (c) harmonic analysis of phase a of load current and reference source current (d) source voltage and current, source voltage and reference current of phase a for PF analysis (continued) (see online version for colours)



(d)

Table 5 Simulation results summary of Case IV

	i_{La}	i_{Sa}^*
THD_i	13.04	12.90
I (A)	26.98	21.38
P (W)	5,013	4,966
Q (var)	3,760	0.297
PF	0.8	1.0
$THD_{va} = 12.62$		

‘This shows that distortion at PCC is mainly due to background distortion (i.e., from grid side) and load side distortion has been compensated. The UPF has compensated reactive power requirement and supplied UPF in this scenario also.’

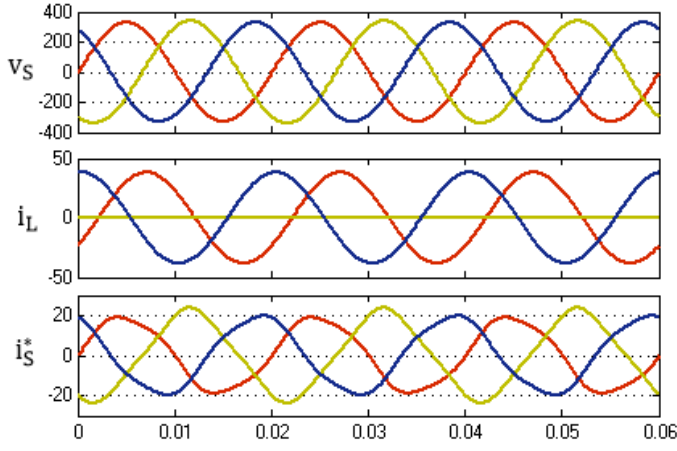
3.5 Case V: ideal supply voltage – unbalanced and undistorted load current (1-phase out)

The only difference between Case I and this case is unbalancing. In this case, unbalancing in load has generated by losing of phase b , i.e., only two phases (phase a and phase c) have supply and third phase is open. In this case, UPF capability of balancing is on test. Figure 10 shows the concerned results.

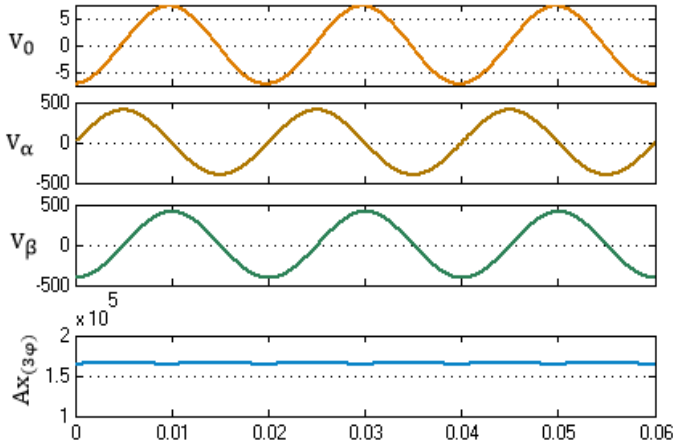
It can be seen from Figure 10(a) that yellow phase is out. Still, algorithm managed to balance the reference current (i_s^*) in three phases equally, although waveforms are not purely sinusoidal and leads to harmonic distortion with THD_i of 8% in each phase.

The effects of supply unbalancing can be seen in v_0 waveform, as shown in Figure 10(b).

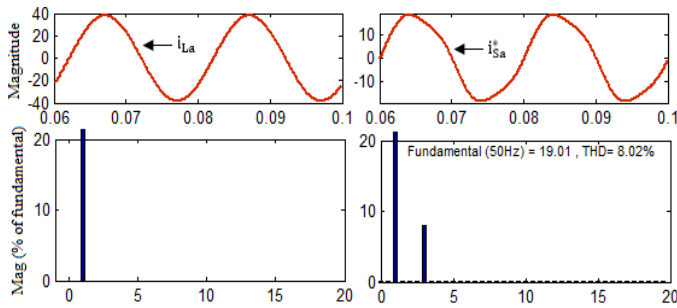
Figure 10 Simulation results for Case V, (a) supply voltages, load currents, and reference source currents (b) 0- α - β coordinates of supply voltages (c) harmonic analysis of phase *a* of load current and reference source current (d) source voltage and current, source voltage and reference current of phase *a* for PF analysis (see online version for colours)



(a)

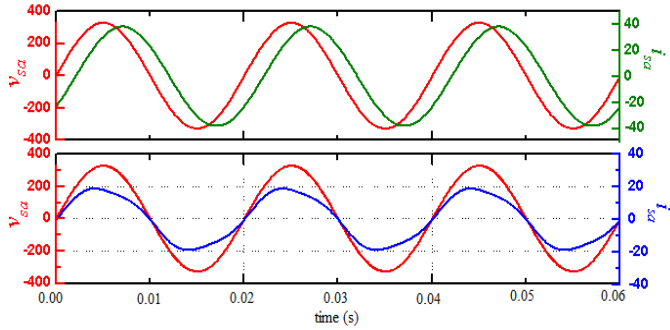


(b)



(c)

Figure 10 Simulation results for Case V, (a) supply voltages, load currents, and reference source currents (b) 0- $\alpha\beta$ coordinates of supply voltages (c) harmonic analysis of phase a of load current and reference source current (d) source voltage and current, source voltage and reference current of phase a for PF analysis (continued) (see online version for colours)



(d)

Equalisation of current in all three phases results a sharp decrease in current RMS value from 27 A to 13.44 A due to distribution of current of two healthy phases into among three phases and further due to compensation of reactive power (shown in Table 6). Negative sign of Q shows reverse flow of reactive power from load to compensator. In such a critical scenario also, the UPF managed to deliver nearly UPF.

Table 6 Simulation results summary of Case V

	i_{La}	i_{Sa}^*
THD_i	0.00	8.02
I (A)	26.98	13.44
P (W)	5,013	3,116
Q (var)	3,760	-196.2
PF	0.8	0.998
$THD_{va} = 0.00$		

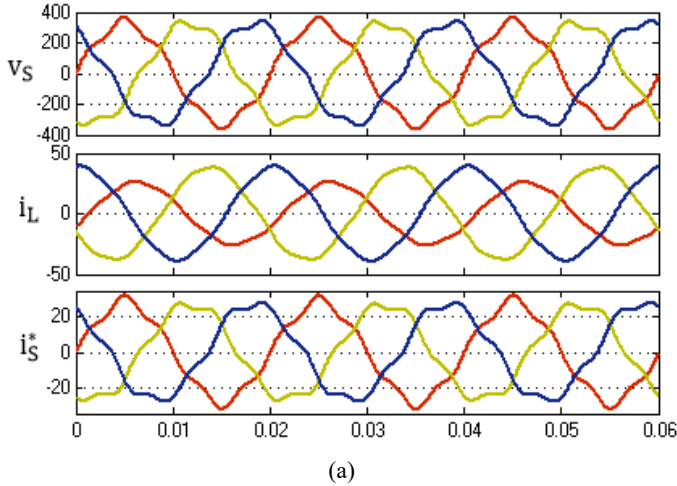
‘So, it is fair to say that the UPF algorithm managed to achieve UPF at PCC even under severe unbalanced load conditions.’

3.6 Case VI: distorted supply voltage (5th harmonic) – unbalanced and undistorted load current

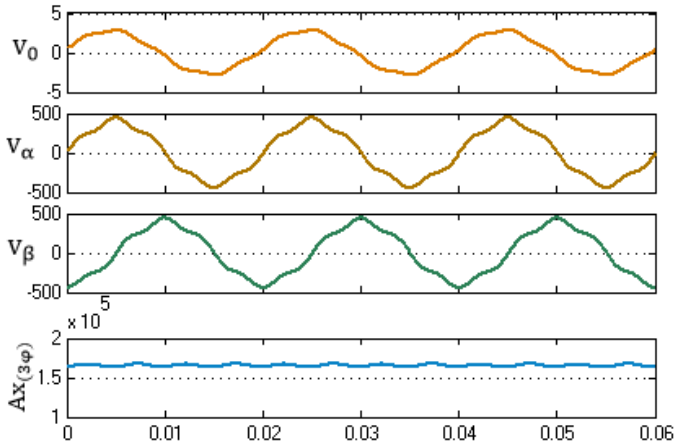
Here, 5th order harmonic is injected in supply voltage as a background distortion. Unbalance in load created by inserting 5 Ω resistance in phase a . Load has no harmonic distortion. Results are shown in Figure 11.

It can be observed from Figure 11(a) that due to insertion of resistance in red phase (phase a), current in this phase has lower magnitude than yellow (phase b) and blue phase (phase c), which show unbalanced load condition. The UPF tries to equalise the current in all the phases as seen from the waveforms of i_s^* in Figure 11(a).

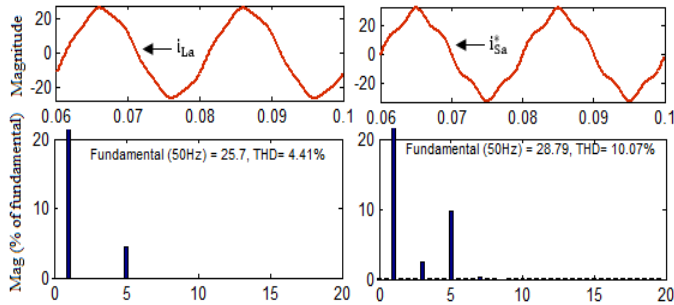
Figure 11 Simulation results for Case VI, (a) supply voltages, load currents, and reference source currents (b) 0- α - β coordinates of supply voltages (c) harmonic analysis of phase *a* of load current and reference source current (d) source voltage and current, source voltage and reference current of phase *a* for PF analysis (see online version for colours)



(a)



(b)



(c)

Figure 11 Simulation results for Case VI, (a) supply voltages, load currents, and reference source currents (b) 0- α - β coordinates of supply voltages (c) harmonic analysis of phase a of load current and reference source current (d) source voltage and current, source voltage and reference current of phase a for PF analysis (continued) (see online version for colours)

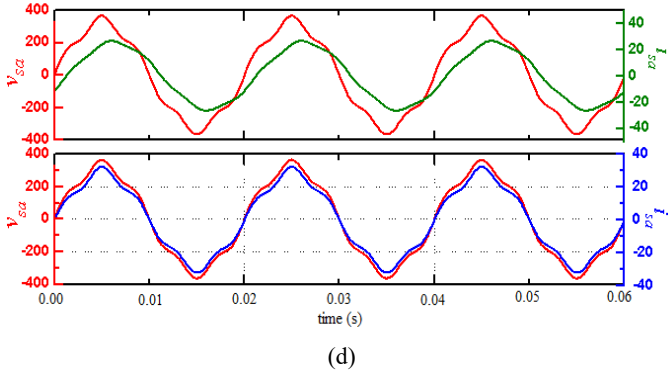


Table 7 illustrate the output parameters for the phase a and phase b . Due to presence of 5th harmonics, source voltage had a THD_v of approximately 9.7% as background distortion, this results distortion in reference current (i_s^*) as 10.07% (THD_i), which is reflection of background distortion 9.7% plus some distortion due to unbalancing in load phases.

Table 7 Simulation results summary of Case VI

	i_{La}	i_{Sa}^*	i_{Lb}	i_{Sb}^*
THD_i	4.41	10.07	3.17	10.39
I (A)	18.88	20.36	26.98	19.38
P (W)	3,928	4,796	5,013	4,502
Q (var)	1,707	27.32	3,760	66.57
PF	0.917	1.0	0.8	0.999
	$THD_{va} = 9.67$		$THD_{vb} = 9.75$	

‘The UPF successfully providing UPF by compensating reactive power requirement, but no improvement has been seen in background distortion coming from grid side.’

3.7 Case VII: distorted supply voltage (5th harmonic) – unbalanced and distorted load current (5th, 7th, 11th harmonics)

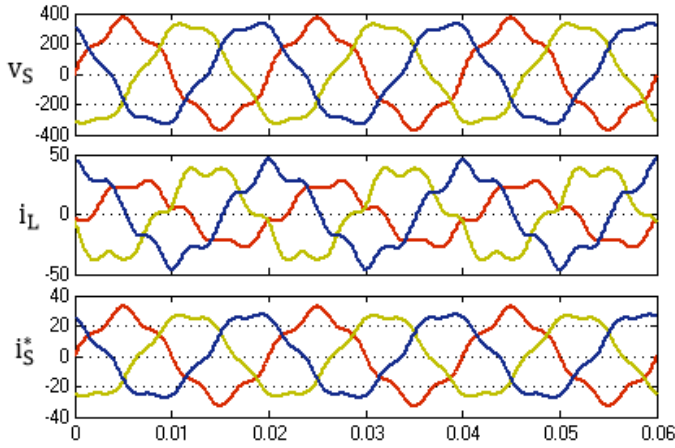
In this case, distortion is created in both, source and load side. Source has 5th harmonics and load has 5th, 7th and 11th harmonics. Unbalancing in load has been created as in the Case VI, i.e., by inserting 5 Ω resistors in phase a . Data related to phase a and phase b has been extracted and corresponding results are shown in Figure 12 and Table 8.

Due to distortion in source voltage and distorted and unbalanced load, voltage THD_v for phases a and b are 12.24% and 8.21%, respectively. Table 8 shows that this voltage THD_v is replicated in reference currents with some addition due to unbalance effect. The UPF has given UPF in this odd situation also. Reactive power requirement has been

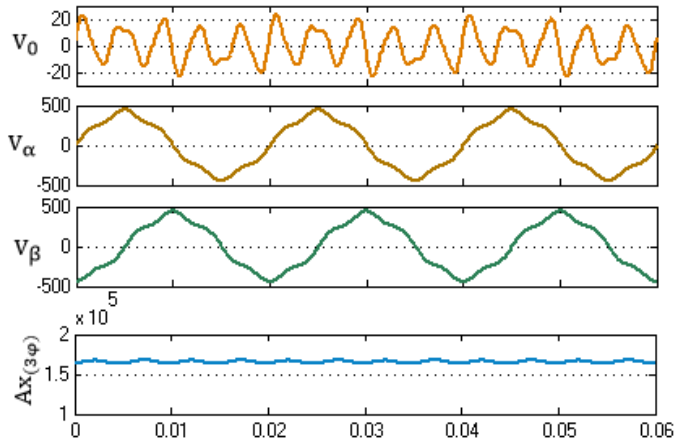
compensated to very low value. The UPF tries to equalise the current in all the three phases which can be seen from Table 8, where load current in phase *a* is 18.18 A but it is compensated at source with 20.32 A and similar for phase *b* from 26.83 A at load side to 19.34 A at source side, which show how UPF controller attempt to equalise current at source side in all three phases to minimise the impact of unbalanced load condition at PCC.

‘Irrespective of disturbances from source side and/or load side, the UPF achieved its target to improve PF as unity and at same time reduce load side unbalancing effects at PCC. However, again there is no improvement on background/grid side distortion.’

Figure 12 Simulation results for Case VII, (a) supply voltages, load currents, and reference source currents (b) 0- $\alpha\beta$ coordinates of supply voltages (c) harmonic analysis of phase *a* of load current and reference source current (d) source voltage and current, source voltage and reference current of phase *a* for PF analysis (see online version for colours)

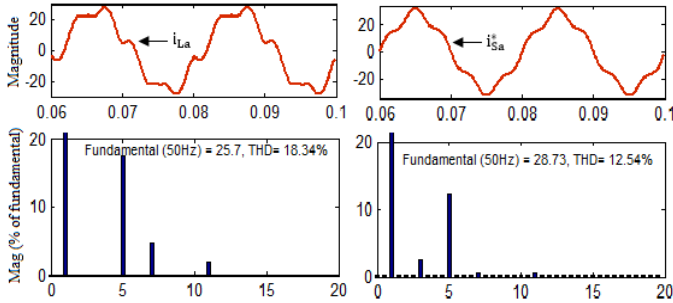


(a)

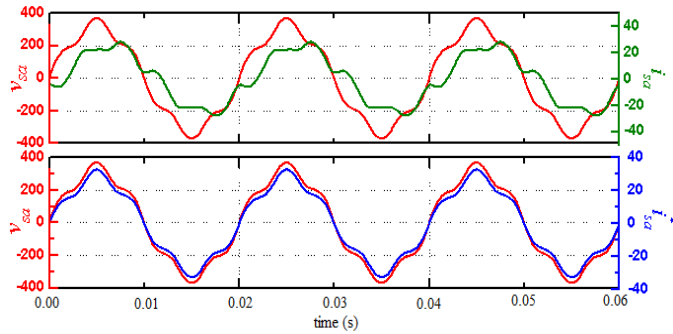


(b)

Figure 12 Simulation results for Case VII, (a) supply voltages, load currents, and reference source currents (b) 0- α - β coordinates of supply voltages (c) harmonic analysis of phase a of load current and reference source current (d) source voltage and current, source voltage and reference current of phase a for PF analysis (continued) (see online version for colours)



(c)



(d)

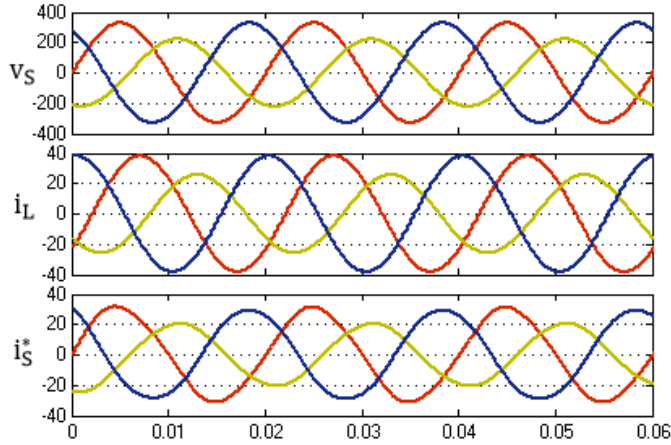
Table 8 Simulation results summary of Case VII

	i_{La}	i_{Sa}^*	i_{Lb}	i_{Sb}^*
THD_i	18.34	12.54	17.83	8.91
I (A)	18.18	20.32	26.83	19.34
P (W)	3,928	4,787	5,013	4,492
Q (var)	1,707	28.06	3760	66.07
PF	0.917	1.0	0.8	0.999
	$THD_{va} = 12.24$		$THD_{vb} = 8.21$	

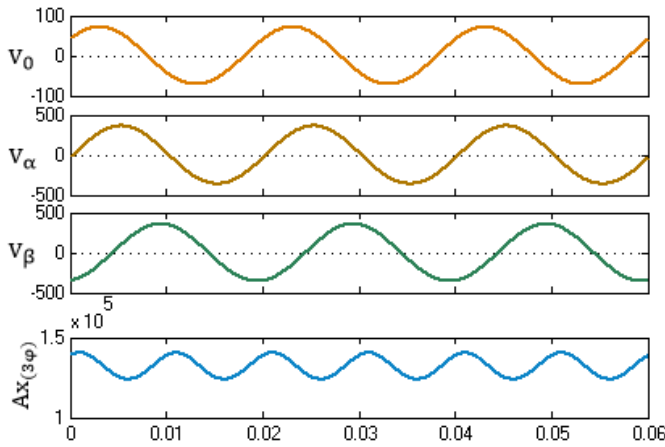
3.8 Case VIII: unbalanced supply voltage – balanced and undistorted load current

This case shows the performance of the UPF control algorithm under unbalanced supply voltage conditions. Load is considered as balanced and without any distortion. Unbalancing in supply is created by inserting additional 5Ω resistor (line impedance is already present) in yellow phase. Results for this case are shown in Figure 13.

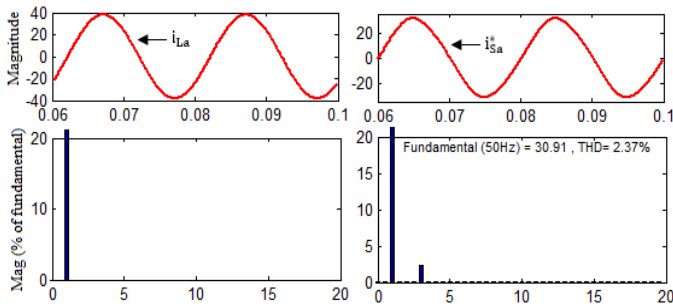
Figure 13 Simulation results for Case VIII, (a) supply voltages, load currents, and reference source currents (b) 0- $\alpha\beta$ coordinates of supply voltages (c) harmonic analysis of phase a of load current and reference source current (d) source voltage and current, source voltage and reference current of phase a for PF analysis (see online version for colours)



(a)

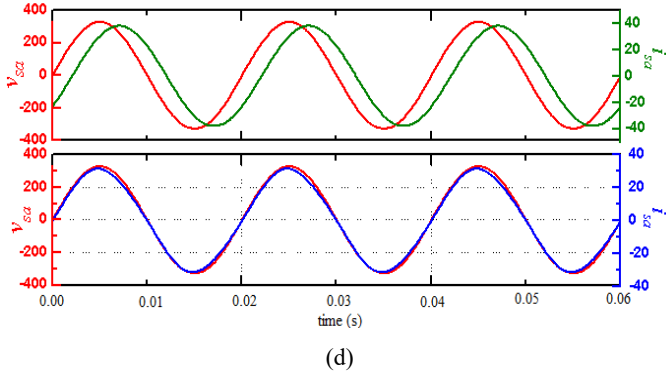


(b)



(c)

Figure 13 Simulation results for Case VIII, (a) supply voltages, load currents, and reference source currents (b) 0- α - β coordinates of supply voltages (c) harmonic analysis of phase a of load current and reference source current (d) source voltage and current, source voltage and reference current of phase a for PF analysis (continued) (see online version for colours)



From Figure 13(a), yellow phase waveform has less magnitude than other two phases. It is due to voltage drop in 5 Ω resistor inserted in this phase. Same can be seen from the waveforms of load current and reference source current. The UPF control theory failed to balance the reference currents although it tried a little as seen from Figure 13(a). As seen in earlier cases also, ‘whenever a disturbance occurs on source side, this algorithm could not observe it and fails to correct it. This fact is confirmed in this case as well.’

Unbalancing effect can be observed in v_0 as shown in Figure 13(b). Same can be seen from $Ax_{(3\phi)}$ waveform. Harmonics distortion (THD_i) in load current is 2.37% and 2.27% in phase a and b , respectively. This distortion is result of UPF efforts to minimise the load current unbalancing effects at PCC. Although it failed, but still PF has been corrected perfectly to nearly unity at PCC, as shown in Figure 13(d). Other parameters are given in Table 9.

Table 9 Simulation results summary of Case VIII

	i_{La}	$i_{S\alpha}^*$	i_{Lb}	i_{Sb}^*
THD_i	0.00	2.37	0.00	2.27
I (A)	26.95	21.85	18.16	14.51
P (W)	5,003	5,071	2,271	2,268
Q (var)	3,752	-94.5	1,703	53.7
PF	0.8	0.999	0.8	0.999
	$THD_{va} = 0.00$		$THD_{vb} = 0.00$	

4 Summary

As discussed in previous section, eight cases have been considered to analyse the capabilities and incapacities of UPF control algorithm applied for DSTATCOM under different source and load conditions. Table 10 summarise the results discussed in previous section.

Table 10 Summary of Cases I to VIII

Cases	Power quality before compensation				Power quality after compensation			
	THD _i (%)	THD _v (%)	PF	Q (var)	THD _i (%)	THD _v (%)	PF	Q (var)
Case I	0	0	0.8	3,752	0	0	1	0.038
Case II	16.08	0	0.8	3,760	3.03	3.04	1	0.432
Case III	3.14	9.75	0.8	3,760	10.04	9.75	1	0.078
Case IV	13.04	9.75	0.8	3,760	12.9	12.62	1	0.297
Case V	0	0	0.8	3,760	8.02	0	0.998	-196.2
Case VI	3.17	9.75	0.8	3,760	10.39	9.75	0.999	66.57
Case VII	17.83	9.75	0.8	3,760	8.91	8.21	0.999	66.07
Case VIII	0	0	0.8	1,703	2.27	0	0.999	53.7

Unbalancing in source voltages
No improvement

Unbalancing in load

Unbalancing in load

Unbalancing in load

Unbalancing in load

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Table 10 provides the comparison of four power quality parameters viz. current THD (THD_i), voltage THD (THD_v), power factor (PF) and reactive power supplied by source (Q). Parameters are compared before and after implementation of selected control theory. By this comparison, capability and incapability of UPF control algorithm to tackle the issues of PF correction, harmonic reduction, reactive power compensation and unbalancing can be evaluated.

In Case I, where ideal supply voltage and balanced and undistorted load currents are taken for study, only issue was poor PF, i.e., 0.8 which required a reactive power of 3,752 var for load. After applying UPF control algorithm, PF improved to unity and reactive power requirement decreased to approximately zero. In Case II, ideal supply voltage with balanced and distorted load current are considered for study. Power quality issues in this case are $THD_i = 16.08\%$, $PF = 0.8$, $Q = 3,760$ var. These issues can be resolved by implementing UPF algorithm, which results THD_i below 5%, UPF and reactive power supplied by source reduced to 0.432 var only. Case II shows that current THD is easily handled by UPF theory and reduced to permissible limit. Case V is also related to ideal supply voltage like as Case I and Case II. In this case, unbalancing is introduced by one phase out. UPF theory has balanced these unbalanced currents and improved the PF to nearly unity. But in attempt to balancing currents, results higher THD_i , which is approximately 8%.

Case I and Case VIII are related to balanced and undistorted load currents, difference lies in supply voltage. Case I has balanced supply voltage and Case VIII has unbalanced supply voltage. Case I has been already discussed in previous paragraph. Output of Case VIII shows, that UPF is unable to balance the source voltages although it has improved the PF to 0.999. This unbalancing also gives rise to current distortion approximately 2.27% as THD_i .

Rest four cases viz. III, IV, VI and VII are related to distorted supply voltages. In these four cases, around 9.75% of THD_v is created as background distortion by injecting 5th harmonic in supply voltage. UPF control theory is not able to correct the voltage distortions in these cases although it is correcting the PF to unity by compensating reactive power requirement of load. This distortion in supply voltage also comes in currents as can be seen in Case III and Case VI where initially THD_i were less but increased after compensation to 10.04% and 10.39% respectively.

5 Conclusions

In this paper, the capabilities and incapacities of UPF control algorithm have been fully analysed at different load and grid conditions. For this purpose, eight different cases have been considered ranging from ideal conditions to highly non-ideal conditions. From all cases, it has been found that UPF control algorithm proved the worth of its name by delivering UPF even under severe non-ideal conditions both from source and load side. Providing UPF is the main goal of this control algorithm and it has been successfully verified. Also, it has been observed that UPF theory has compensated reactive power requirement of the load and therefore, source has to supply only active power, this helps to reduce nonlinearity effects of power electronic loads and reduce source current magnitude. This reduction in source current will help to reduce losses in cable and distribution transformer, which ultimately improve reliability and life-time of whole distribution system.

In case of unbalanced load condition, the UPF theory tried to equalise the current in all the phases and achieved it up to a satisfactory level. In case of distortion in load side, this theory has been successfully compensated the distortion well below the IEEE-519 standard limits, i.e., 5% of THD_v . However, in case of distortion coming from source side, the UPF theory failed to correct it. It is due to the fact that UPF sees the problems occurring after the PCC and correct them, but it does not notice the problems related to source side and as a result replicate problem of source in the reference current.

Based on this discussion, it can be concluded that UPF control algorithm is able to provide UPF in any abnormal condition both from source and or load side by compensating reactive power requirements from load. This can reduce the unbalanced load effects at PCC. It also minimises the harmonic distortion if harmonics are generated from load side only.

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